

On the Statistics of the Sum of Correlated Generalized- K RVs

ICC'10

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Room: Auditorium 2

14:00-15:45

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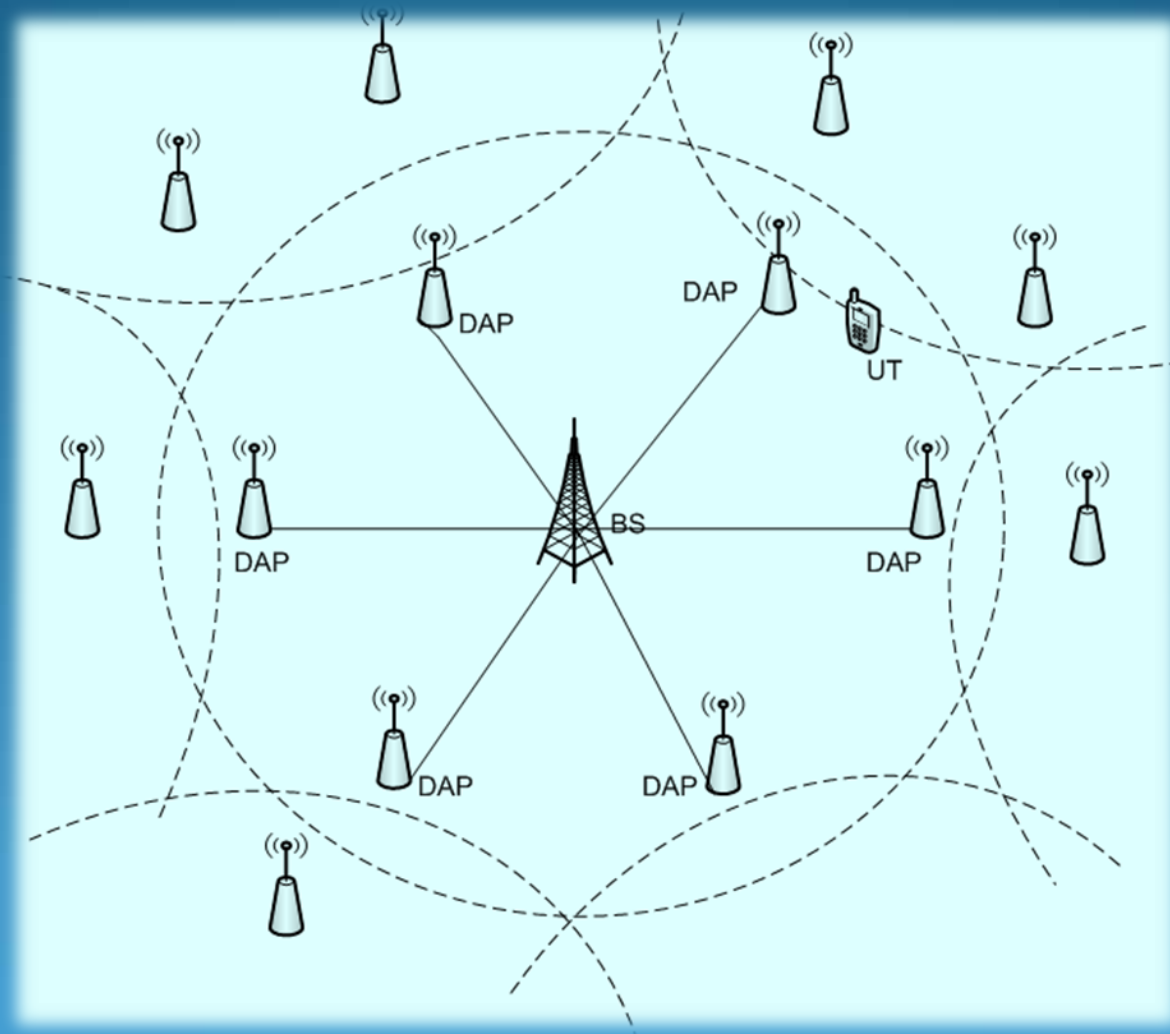
Presentation Outline

- Introduction
- Generalized- K Composite Fading Model
- Related Work on the Statistics of Correlated Generalized- K Random Variables (RVs)
- Amount of Fading (AF) Expressions
- Results
- Conclusions

Introduction

- Composite channel models take place in wireless communications (multi-path fading plus shadowing), radar cross section scattering of targets, reverberation in sonar, etc.
- Modeling such a phenomenon plays an important role in the design and performance analysis for such channels.
- In wireless communications, emerging systems such as coordinated multipoint transmission and reception (CoMP) systems (network MIMO) including distributed antenna systems, and cooperative relay networks require such modeling.
- Shadowing correlations are typical in such wireless geographically distributed systems.

Ex. A multi-cell DAS



Models of Composite Fading

Multipath	Shadowing	Composite	Complexity	Appeared
Rayleigh	lognormal	Suzuki	No closed-form expression	Suzuki (1979)
Nakagami	lognormal	Gamma-lognormal	No-closed form expression	Lewinsky (1983) Stuber (2001)
Rayleigh	Gamma	Exponential-Gamma	Closed-form but limited	Jakeman (1976) Abdi (1998)
Nakagami	Gamma	Gamma-Gamma	Closed-form	Lewinsky(1983) Shankar (04)

The Generalized-K Model

- Using the Nakagami model for multipath fading and the Gamma model for shadowing results in the Gamma-Gamma (generalized-K) model

Multipath fading

shadowing

$$p_{\gamma/\Omega}(x) = \frac{1}{\Gamma(m_m)} \left(\frac{m_m}{\Omega} \right)^{m_m} x^{m_m-1} \exp\left\{ -\frac{m_m x}{\Omega} \right\}, \quad x > 0, m_m \geq 0.5$$

and

$$p_{\Omega}(y) = \frac{1}{\Gamma(m_s)} \left(\frac{m_s}{\Omega_0} \right)^{m_s} y^{m_s-1} \exp\left\{ -\frac{m_s}{\Omega_0} y \right\}, \quad y \geq 0, m_s > 0$$

After averaging,

$$p_{\gamma}(x) = \int_0^{\infty} p(x/\Omega) p_{\Omega}(y) dy$$

$$p_{\gamma}(x) = \frac{2}{\Gamma(m_s)\Gamma(m_m)} (b)^{m_s+m_m} x^{\left(\frac{m_s+m_m}{2}\right)-1} K_{m_s-m_m}(b\sqrt{x}), \quad x \geq 0, m_m \geq 0.5, m_s > 0$$

Where m_m , m_s are the multipath fading and shadowing parameters,

$K_{m_s-m_m}(\cdot)$ is the Bessel function of the second type and $b = 2\sqrt{\frac{m_m m_s}{\Omega_0}}$ where Ω_0 is the mean of the local power.

The Generalized- K Model (Contd)

- The generalized- K (GK) PDF, being the PDF of the product of two independent Gamma RVs, belongs to the family of Fox H -function PDF family.
- The PDF of the product of N independent H -function RVs is another H -function PDF as well as the PDF of the quotient of two independent H -function RVs [Springer, 1979]. However, not for the sum.
- However, no closed-form expression for the PDF of the product, quotient, or the sum of correlated H -function RVs is known. Otherwise, the PDF of the sum of correlated Rayleigh, Nakagami, etc. would have been known.

Related Work On the Statistics of Correlated GK RVs

- The simple case where the shadowing components are fully correlated (collocated antenna systems) was considered in [Shankar, 2006] where it was shown that the sum follows another GK distribution with $m_{m,sum} = Nm_m$ and $m_{s,sum} = m_s$.
- Infinite series expressions for the PDF, the CDF, and the CHF of the joint distribution were derived in [P. Bithas, et.al.] and the performance of maximal ratio combining (MRC) and equal gain combining for a dual diversity combiner are studied.

Total correlation coefficient between two GK RVs

The sum of N correlated generalized- K RVs can be expressed as

$$\xi = z_1 w_1 + z_2 w_2 + \dots + z_N w_N$$

where z_i and w_i denote the multipath fading and shadowing RVs, respectively. In general, while the multipath and shadowing components are independent, correlation among the z_i 's and correlation among the w_i 's may exist in certain propagation scenarios. So,

$$\rho_{i,j} = \frac{E[z_i w_i z_j w_j] - E[z_i w_i] E[z_j w_j]}{\sigma_{z_i w_i} \sigma_{z_j w_j}}, \quad i, j = 1, \dots, N.$$

However, we may write $E[z_i w_i z_j w_j] = E[z_i z_j] E[w_i w_j]$

Since multipath fading is independent from shadowing

Furthermore, using the composite fading multiplicative model, the correlation coefficient can be expressed as

$$\rho_{i,j} = \frac{\rho_{z_i, z_j} \sqrt{m_{s,i} m_{s,j}} + \rho_{w_i, w_j} \sqrt{m_{m,i} m_{m,j}} + \rho_{z_i, z_j} \rho_{w_i, w_j}}{\sqrt{m_{m,i} + m_{s,i} + 1} \sqrt{m_{m,j} + m_{s,j} + 1}}$$

The expression simplifies for the identically distributed case to (i.d.) Case to

$$\rho_{i,j} = \frac{\rho_{z_i, z_j} m_s + \rho_{w_i, w_j} m_m + \rho_{z_i, z_j} \rho_{w_i, w_j}}{m_m + m_s + 1}$$

The Amount of Fading (AF) expressions

The amount of fading (AF) is defined as the ratio of the variance to the squared mean

$$AF = \frac{\text{var}(\xi)}{[E(\xi)]^2}$$

The AF was introduced to quantify the fading severity; however, it has been shown later that it can be related to some performance measures like the symbol error rate [B. Holter, and E. Oien, 2005] and the ergodic capacity [Y. Li and Kishore, 2008].

The AF for the sum of correlated GK RVs can be expressed as

$$AF_{\xi} = \frac{\sum_{i=1}^N AF_i \Omega_{0,i}^2 + \sum_{i=1}^N \sum_{j=1, i \neq j}^N \rho_{i,j} \sqrt{AF_i} \sqrt{AF_j} \Omega_{0,i} \Omega_{0,j}}{\left(\sum_{i=1}^N \Omega_{0,i} \right)^2}$$

For i.i.d. case and a constant correlation model

$$AF_{\xi}^{i.i.d.e.c} = \frac{1-\rho_m}{Nm_m} + \frac{1-\rho_s}{Nm_s} + \frac{1-\rho_m\rho_s}{Nm_m m_s} + \frac{\rho_m}{m_m} + \frac{\rho_s}{m_s} + \frac{\rho_m\rho_s}{m_m m_s}$$

Cont.....

Typically, for geographically spaced antennas, only shadowing correlations are significant; hence the AF expression reduces to

$$AF^{i.d.e.c} = \frac{1}{Nm_m} + \frac{1-\rho_s}{Nm_s} + \frac{1}{Nm_m m_s} + \frac{\rho_s}{m_s}$$

Clearly $AF_{N \rightarrow \infty}^{i.d.e.c} = \frac{\rho_s}{m_s}$ as diversity does not combat the correlated part.

Negative correlation may take place in some propagation [K. Butterworth, et.al. 1997, E. Perhia, et.al, 2001].

For $N=2$, the effect of shadowing correlation vanishes for $\rho_s=-1$. In general, it can be shown that the effect of shadowing correlation vanishes for $\rho_s=-1/(N-1)$ for $N>1$.

Application to Capacity of DAS Systems

The effect of shadowing correlations on macrodiversity (DAS) can be studied using the following approximation [Y. Li and Kishore, 2008]

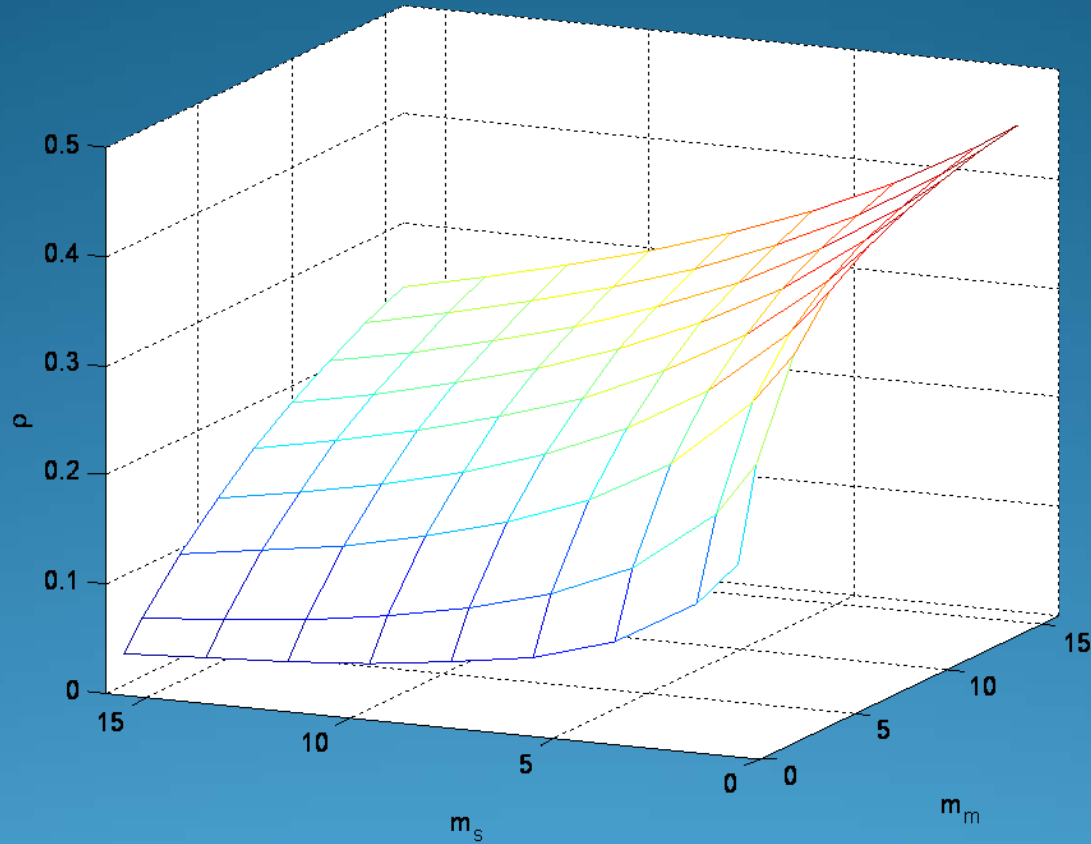
The ergodic capacity, after MRC, can be approximated, (for $AF < 0.5$), as

$$C \approx \log_2(1 + SNR) - \frac{\log_2}{2} AF \left(\frac{SNR}{SNR + 1} \right)^2$$

$$As, \xrightarrow{N \rightarrow \infty} C \approx \log_2(1 + SNR) - \frac{\log_2}{2} \frac{\rho_s}{m_s} \left(\frac{SNR}{SNR + 1} \right)^2 \xrightarrow{SNR \rightarrow \infty} C_{loss} \approx \frac{\log_2}{2} \frac{\rho_s}{m_s} \text{ b/s/Hz}$$

Hence, the asymptotic ergodic capacity loss due to correlated shadowing can be predicted for different values of m_s and ρ_s as far as $\frac{\rho_s}{m_s} \leq 0.5$.

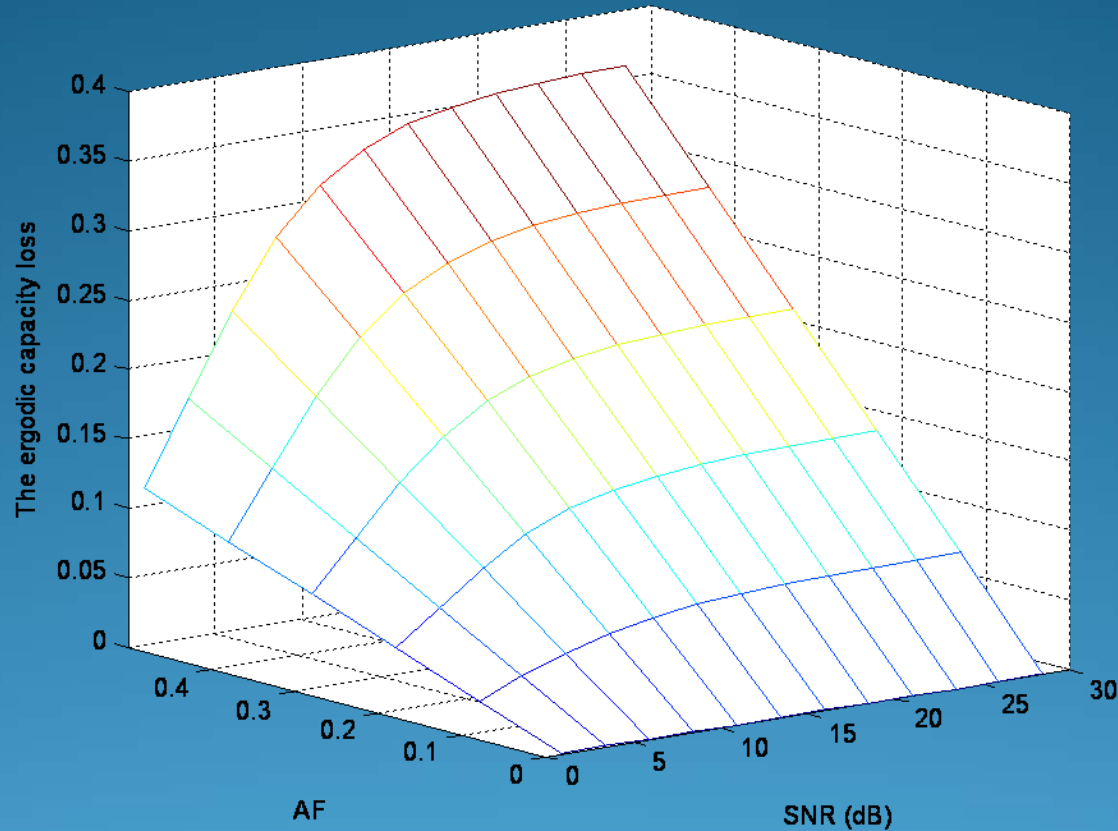
Results



The plot of the correlation coefficient between two GK RVs for $\rho_s=0.5$.

The correlation coefficient decreases as m increases (less shadowing) and as m decreases where the multipath components dominate.

Results



The plot for the ergodic capacity loss versus the AF and the SNR.

The asymptotic ergodic capacity loss for typical shadowing correlation scenarios (where $\rho_s=0.5$) is bounded by less than 0.4 b/s/Hz.

Conclusions and Future Work

- The generalized- K composite fading model is more tractable than lognormal-based model. However, some challenges are there.
- The correlation coefficient between two generalized- K RVs is derived and subsequent expressions for the AF are presented.
- Moreover, the effect of negative correlations is highlighted.
- The effect of shadowing correlations on the performance of MRC receivers can be predicted.
- The formulation can be extended to MIMO scenario using the channel matrix Frobenius norm.

References

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- [2] P. M. Shankar, "Performance analysis of diversity combining algorithms in shadowed fading channels," *Wireless Personal Communications.*, v. 37, no. 2, Apr. 2006.
- [3] Bithas, P.S. and Sagias, N.C. and Mathiopoulos, P.T., "The bivariate generalized- K KG distribution and its application to diversity receivers," *IEEE Transactions on Communications.*, v. 57, no. 9, Sep. 2009.
- [4] Li, Y. and Kishore, S., "Diversity factor-based capacity asymptotic approximations of MRC reception in Rayleigh fading channels," *IEEE Transactions on Communications.*, v. 56, no. 6, June 2008.
- [5] Holter, B.; Oien, G.E.; , "On the amount of fading in MIMO diversity systems," *IEEE Transactions on Wireless Communications*, vol.4, no.5, pp. 2498- 2507, Sept. 2005.

Appendix: the Fox H -function

Definition: The H -function can be defined as

$$\mathbf{H}_{p,q}^{m,n} \left[z \middle| \begin{smallmatrix} (a_1, \alpha_1), \dots, (a_p, \alpha_p) \\ (b_1, \beta_1), \dots, (b_q, \beta_q) \end{smallmatrix} \right] = \frac{1}{2\pi i} \int_C \frac{\prod_{j=1}^m \Gamma(b_j - \beta_j s) \prod_{j=1}^n \Gamma(1 - a_j + \alpha_j s)}{\prod_{j=m+1}^q \Gamma(1 - b_j + \beta_j s) \prod_{j=n+1}^p \Gamma(a_j - \alpha_j s)} z^s ds, \quad (\text{A.1})$$

where $0 \leq m \leq q$, $0 \leq n \leq p$, $\alpha_j > 0$, $\beta_j > 0$. and a_j ($j = 1, 2, \dots, p$) and b_j ($j = 1, 2, \dots, q$) are complex numbers such that no pole of $\Gamma(b_j - \beta_j s)$ for $j = 1, 2, \dots, m$ coincides with any pole of $\Gamma(1 - a_j + \alpha_j s)$ for $j = 1, 2, \dots, n$.

Many of the so-called special functions are special cases of the H -function, including Gauss and confluent hypergeometric functions, MacRobert's E-function, Meijer's function; and Bessel functions. The Meijer's function can be expressed as

$$\mathbf{G}_{p,q}^{m,n} \left[x \middle| \begin{smallmatrix} a_1, \dots, a_p \\ b_1, \dots, b_q \end{smallmatrix} \right] = \mathbf{H}_{p,q}^{m,n} \left[z \middle| \begin{smallmatrix} (a_1, 1), \dots, (a_p, 1) \\ (b_1, 1), \dots, (b_q, 1) \end{smallmatrix} \right]. \quad (\text{A.2})$$

Definition: The H -function distribution can be expressed as

$$p(x) = k \mathbf{H}_{p,q}^{m,n} \left[cx \middle| \begin{smallmatrix} (a_1, \alpha_1), \dots, (a_p, \alpha_p) \\ (b_1, \beta_1), \dots, (b_q, \beta_q) \end{smallmatrix} \right], \quad x > 0, \quad (\text{A.3})$$

where k and c are the parameters of the distribution such that $\int_0^\infty p(x) dx = 1$.

Questions