

Interference Management in Underlay Spectrum Sharing Using Indirect Power Control Signalling

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Abstract—In this paper, we propose an interference management method for underlay spectrum sharing and evaluate its performance. In this method the secondary service has been facilitated by granting passive access to the power control signalling transmitted by the primary network base station. To exploit both slow shadowing and fast fading, the proposed method performs secondary service power management in two phases, each in different time-scales: *rate optimal power allocation phase* and *interference reduction power adjustment phase*. In the longer time-scale, *rate optimal power allocation phase* adaptively allocates the secondary transmit power exploiting the medium-scale channel variations (shadowing effect) of the secondary channel to maximize its capacity. In the shorter time-scale, *interference reduction power adjustment phase* exploits the power control commands transmitted in the primary network to adaptively adjust secondary service transmission power for reducing the effects of the secondary service transmission on the Quality-of-Service (QoS) of the primary service network. The main advantage of this method which is referred to as Adaptive Multiple Time-Scale Power Allocation (AMTPA) is that it does not require direct signaling between the two systems. We further present AMTPA analytical performance evaluation results. Practical considerations are also presented regarding the primary network requirements and its power control feasibility after adopting AMTPA in the secondary network. Extensive simulation results indicate significant improvement in the system performance by using AMTPA. Using simulations we also show how one can set AMTPA parameters so that a certain level of QoS in the primary network is satisfied.

Index Terms—Collision probability, DS-CDMA, interference management, underlay spectrum access, spectrum sharing.

I. INTRODUCTION

IN spectrum sharing a *secondary service* is able to access the white spaces in a frequency band which is formally allocated to the *primary service* [1]. White spaces are those parts of the spectrum allocated to the primary user which are under-utilized in some particular times and specific locations. Various schemes are proposed in the literature for spectrum sharing (see e.g., [2] and references therein). Our focus is on the underlay spectrum sharing [3]. In underlay spectrum sharing, the secondary service can always access to the spectrum subject to an interference threshold constraints at the primary service receiver [2], [3].

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To satisfy the interference threshold constraint, in underlay spectrum sharing, Channel State Information (CSI) between the secondary transmitter and receiver (*secondary-channel*), and between the primary receiver and the secondary transmitter (*cross-channel*) are required (see, e.g., [4], [5], [6]).

In practice, the cross-channel CSI may be feedback directly by the primary service to the secondary transmitter, measured by the secondary transmitter, or indirectly through a higher layer band manager which coordinate spectrum access [1]. However, the aforementioned scenarios are difficult to be implemented since the primary network has been already deployed and not been designed to provide direct signalling to other networks. To tackle this, different methods have been presented in the related literature.

In [7], [8] a low bit rate signaling channel is considered between the primary and the secondary systems for communicating the channel side information. For measuring the channel gains, pilot signals are transmitted by both systems. However, the pilot transmission and its corresponding signaling channel impose a system overhead that may not be acceptable to the primary network. To resolve this, in [9] a scheme is proposed by which the secondary service estimates the power spectrum of the interference using a multi-taper method. Utilizing a large number of sensors is also proposed in [10] to sniff the radio frequency environment. Such method encounters practical concerns in a civilian wireless network.

Recently approaches based on exploiting the channel feedback in the primary network by the secondary service, have been proposed in the related literature. In such approaches, the feedback is utilized by the secondary network to manage the impact of spectrum sharing on the primary service as well as to enhance the secondary service performance. In [11], a spectrum sharing scenario has been proposed in which the secondary service overhears and exploits the Automatic Repeat Request (ARQ)s signaling in the primary service and accordingly adjusts its access to the spectrum. It is also shown in [11] that this passive overhearing guarantees primary service target rate. Similarly in [12], a feedback-based access scheme is suggested in which the secondary service is allowed to overhear the primary service ARQ channel. If ACK is observed, the secondary service will then access the channel; otherwise access is deferred to reduce almost certain collision outcome between the secondary and primary transmissions. Only one primary transceiver and secondary transceiver is considered and the analysis is conducted for an erasure channel in which primary channels erase probability is changed due to secondary service activity.

Furthermore, in [13] for a system model including a pair

of secondary and primary services in which the secondary service has access to the quantized channel state information of the primary channel, i.e., the channel between the primary transmitter and receiver, they then formulate an optimization problem to maximize the secondary service capacity subject to primary service rate constraint. Having access to this information the secondary service is then able to speculate the rate reduction at the primary service due to its transmission.

In [14] a method has been proposed in which the secondary service is provided with passive access to one information bit that indicates the status of the outage constraint at the primary receiver. However, only one primary transceiver is considered and impact of shadowing is also ignored.

To tackle the cross-channel CSI issue, here we propose an indirect power control signalling method namely Adaptive Multiple Time-Scale Power Allocation (AMTPA) for underlay spectrum sharing over uplink of Direct Sequence Code Division Multiple Access (DS-CDMA) primary networks. In AMTPA power control signalling in the primary system is utilized by the secondary service to adjust its transmit power based on the primary network actual status; thus to manage the interference. Orthogonal frequency division multiple access (OFDM) is utilized by the secondary service to access the available spectrum.

A. DS-CDMA/OFDM Spectrum Sharing

DS-CDMA is the dominant air interface technique 3G mobile communications and some Wireless Local Area Network (WLAN) technologies. Therefore, the spectrum sharing over existing DS-CDMA-based networks is anticipated to be one of the spectrum sharing applications in the near future.

From the primary service perspective, the spreading characteristic of DS-CDMA makes it more robust to the narrow-band interference created by OFDM secondary users. From the secondary service perspective, OFDM provides the required flexibility to the secondary service to access separate appropriate portions of the spectrum. Since the wireless channel is frequency selective, appropriate portions (i.e., sub-channels) might be found for secondary transmission which impose minimum interference on the primary service, and at the same time achieve as much capacity as possible in the secondary system. For instance consider the case that there are some sub-channels with a deep fade over the link between the secondary transmitter and the primary receiver, and a good channel condition over the link between the secondary transmitter and receiver. Such sub-channels, if selected for secondary transmission, result in a high secondary service capacity, and at the same time impose low narrow-band interference over the primary network. The impact of this imposed narrow-band interference is further decreased in the primary network because of DS-CDMA technology. Therefore, DS-CDMA/OFDM combination provides a new degree of freedom by enabling the secondary service to adaptively select appropriate sub-channels.

B. The Proposed Approach

To utilize AMTPA in an existing network, the only requirement is an *indirect inter-system cooperation* between the

primary network and secondary service. Such indirect inter-system cooperation, in its simplest form, facilitates secondary service access to the corresponding power control commands of a particular user of the primary service which we refer to as the ‘probe user’. In fact, the main advantage of the AMTPA is that the primary system does not require to be modified to support AMPTA. However, the scheme does require a degree of cooperation and co-design to enable the secondary users to have access to the corresponding signaling channels to receive the power control commands in the primary system. Comparing to establishing a new direct signaling channel between the primary and secondary systems, AMTPA requires much lower level of cooperation between the two system.

Our focus is spectrum sharing in the uplink of the primary network. That is because most of the modern data applications are asymmetric, i.e., the amount of downlink communications is much higher than that of the uplink. Therefore, for spectrum sharing over 2G and 3G systems, the uplink spectrum is most likely under-utilized and can be used for spectrum sharing.

To maintain a certain level of wireless link quality, in DS-CDMA systems a power control mechanism is adopted [15]. The transmitter then adjusts its transmit power based on the received power control commands to maintain the link quality. The base station (BS) of the primary service routinely reacts to degrading the wireless link quality by sending suitable power control commands to the affected primary service users. In case of underlay spectrum sharing, the interference caused by the secondary users, if not the only reason, is one of the crucial factors in degrading the link quality in the primary network. Therefore, the power control commands contain information reflecting the interference level in the primary network.

Through listening to power control commands in AMTPA, the secondary service users can adjust their transmit powers to satisfy the link quality constrain in the primary network. The secondary service employs underlay spectrum sharing for accessing a portion of the uplink spectrum, namely *sub-channel*. A two level time hierarchy consisting of “frames” and “slots” is considered in which, each frame is divided into a number of slots. Secondary service transmission is managed in two phases: *rate optimal power allocation phase* and *interference reduction power adjustment phase*, performed in the frame and slot time-scales, respectively.

The rate optimal power allocation phase adaptively allocates the optimal transmit power using the instantaneous status of the secondary channel with the objective of maximizing the capacity of the secondary service without considering the impact of secondary transmission on the link quality in the primary network. Utilizing the power control commands of the primary network, the allocated power is then adjusted in the interference reduction power adjustment phase, so that the required link quality in the primary network is satisfied.

Power adjustment is conducted through increasing/decreasing transmission power with fixed step size Δ_s Watts. Modified versions of AMTPA are also presented to further improve the system performance. We also propose a scheme for selecting the probe user. Interestingly, we show that the probe user can be selected by the primary service based-station based on the channel power gains among the BS and the primary users, and the eigenvalues of the primary users channel matrix.

We also discuss practical considerations and the feasibility of the power control in the primary network when AMTPA is adopted by the secondary network. This paper provides an instance of the approaches in which the inherent structure of the primary network can be exploited by the secondary service. This is significantly important since using such approaches minimum modifications in the primary network is required to facilitate spectrum sharing.

Simulations are performed to study different performance aspects of the proposed method. To the authors' best knowledge, there is no other fundamentally similar approach to AMTPA; therefore, in our simulations, among other things, we compare the performance of AMTPA with the existing performance bounds in the related literature. The simulation results indicate improvement in the system collision performance by using AMTPA. Using simulations we also show how one can set the AMTPA system parameters so that a certain level of quality-of-service (QoS), e.g., collision probability in the primary network is satisfied.

C. Contributions

Our main objective in this paper is to show how the secondary service can exploit one bit power control command at the primary service to guarantee the primary service quality-of-service in terms of outage probability and power control feasibility. The contributions of this paper are as follows:

- A new secondary service power allocation is proposed in which an indirect signaling based on one bit power control commands in the primary network is utilized to adjust the transmit power in the secondary network.
- The proposed multi time scale power allocation exploits both slow shadowing and fast fading to maximize the achievable rate of the secondary users.
- A comprehensive performance evaluation has been developed for the proposed power allocation in which errors that might happen in overhearing the primary service power control command are appropriately taken into account and the associated negative impacts on the performance of the primary and secondary service are analyzed.

D. Organization of the Paper

In Section II, the system model is presented. In Section III, we present and model the underlay spectrum sharing in DS-CDMA primary network. Then, in Section IV we present AMTPA method and investigate its performance. Practical considerations regarding the primary network, the probe user and power control feasibility in the primary network are also discussed in Section V. Simulation results are presented in Section VI, followed by conclusions in Section VII.

II. SYSTEM MODEL

Two services try to access the spectrum band: *Primary Service* and *Secondary Service*. The concerned frequency band is licensed to the primary service which we refer to as 'spectrum'. The primary network is a cellular DS-CDMA system with processing gain G (see, e.g., [16]). The secondary service does not have the spectrum license, but may access the

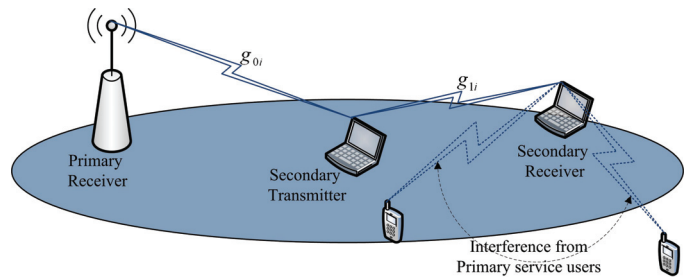


Fig. 1. System schematic for a sub-channel i .

spectrum by adopting underlay spectrum sharing. Subscripts s and p refer to the secondary service and the primary service, respectively. We define $I_{p|s}$ ($I_{s|p}$) as the interference created by the activity of the secondary service (primary service) at the primary (secondary) receiver.

We consider a frequency-selective wireless channel with additive white Gaussian noise (AWGN). The white noise power spectral density is N_0 . From the secondary service viewpoint, the channel is divided into N flat fading B_c Hz sub-channels, where B_c is much smaller than the channel coherence bandwidth. Therefore, the total bandwidth is $B = NB_c$. Sub-channels are indexed by i , $i = 1, 2, \dots, N$. The wireless channel also experiences a log-normal shadowing with zero mean and standard deviation of σ_{dB} dB. The coherence bandwidth of shadowing is usually much larger than that of flat fading; therefore, we assume that shadowing is constant for the whole bandwidth.

An schematic of the considered system in this paper is depicted in Fig. 1. For sub-channel i , g_{0i} and g_{1i} in Fig. 1 denote the instantaneous gains of sub-channel i from the secondary transmitter to the primary receiver (i.e., *cross-sub-channel*) and the secondary receiver (i.e., *secondary-sub-channel*), respectively. Both g_{0i} and g_{1i} are assumed to be independent random variables with probability density functions (pdf), $f_{0i}(g_{0i})$, $f_{1i}(g_{1i})$, respectively. Both g_{0i} and g_{1i} are i.i.d. for $i = 1, 2, \dots, N$.

Time is divided into frames, indexed by n , each with T_{frame} seconds duration. Frame duration T_{frame} is much smaller than that of the shadowing coherence time; therefore, during each frame, the shadowing gain is constant. Each frame consists of N_p slots indexed by k , $k = 1, 2, \dots, N_p$ each with T_{slot} seconds duration. The duration of slots T_{slot} is much smaller than that of the coherence time of channel fading.

Link level performance in the primary network is a function of the corresponding signal to interference plus noise ratio (SINR). To achieve reliable communication, the SINR, or equivalently the bit energy to the interference spectral density (E_b/I_0) should be above a certain threshold, namely ρ . For maintaining the targeted E_b/I_0 as the link level performance threshold in DS-CDMA systems, a power control mechanism is usually adopted [16]. Power control commands are communicated to the transmitter in an appropriate time-scale. Based on the received power control command, the transmitter then adjusts its transmit power to maintain targeted threshold for the required quality of service. Therefore, E_b/I_0 for all users in the BS coverage area is managed to be larger than or equal to ρ , which is a system parameter.

Here, we assume a fixed-step power control (FSPC) algorithm. FSPC is widely used in cellular DS-CDMA standards

such as UMTS and IS-95 systems due to its simplicity and stability [17], [18]. The time-scale of FSPC is T_{slot} , i.e., the power control commands are transmitted each T_{slot} seconds. In FSPC, the BS measures the corresponding received E_b/I_0 to an active mobile user and then compares it with the threshold, ρ . A one-bit power control command, $\Gamma_{pm}(n, k) \in \{0, 1\}$, is then sent to the mobile station m in the k th slot of the n th frame. If $\Gamma_{pm}(n, k) = 1$ ($\Gamma_{pm}(n, k) = 0$) it indicates that the received E_b/I_0 is less (greater) than ρ thus the mobile should increase (decrease) its transmit power by Δ_p Watts.

III. UNDERLAY SPECTRUM SHARING IN DS-CDMA PRIMARY NETWORK

Let M be the number of active primary users in the uplink of a DS-CDMA cellular network indexed by m , $m = 1, 2, \dots, M$. Without spectrum sharing, E_b/I_0 requirement for this system is satisfied if [15]

$$E_b/I_0 = \frac{GQ_m}{\sum_{j \neq m} Q_j + N_0B} \geq \rho, \quad m = 1, 2, \dots, M, \quad (1)$$

where G is the processing gain of DS-CDMA system, Q_j is the received power corresponding to user j at the front-end of the primary service receiver (i.e., BS), and ρ is the E_b/I_0 threshold.

As the performance measure, we consider the outage probability which is defined as

$$P_{out} \triangleq \Pr \left\{ \frac{E_b}{I_0} < \rho \right\}, \quad (2)$$

where $\Pr\{A\}$ is the probability of occurrence of event A . In the ideal case, where $P_{out} = 0$, the maximum number of active users in the primary network is [15]

$$M_0 = \frac{G}{\rho} + 1. \quad (3)$$

If $P_{out} > 0$, then the number of active users in the network coverage area may exceed M_0 .

For a large number of users in DS-CDMA networks, invoking the Central Limit Theorem justifies the Gaussian approximation for the interference process [16]. Using second-order statistics, it is shown that the interference process is also white [15]. In practice, the above argument is shown to be valid for a moderate number of users (e.g., 10-30) [19], [20]. Moreover, due to the fundamental limitation of the conventional DS-CDMA receivers (see e.g., [16]) the interference level in the receiver should not be increased beyond a threshold which is the product of the background noise power spectral density and a constant coefficient, which is a system parameter. Therefore, in DS-CDMA networks, the total received interference caused by other active users in the network coverage area is simply modelled as an additive white Gaussian noise. Moreover, the power spectral density of the received interference is equal to the product of the background noise power spectral density and a factor $(K - 1)$ [15], [21]. In the following we obtain K .

In DS-CDMA network for each transmitter, a silence period detector is usually utilized to exploit the silence periods by stopping transmission during such periods. This technique reduces the interference thus results in a capacity gain. To

model the user's activity in DS-CDMA network, users' activity factor, ν , is defined that is the fraction of time in which the user is transmitting (see, e.g., [15], [22]). The more activity in the primary network, the higher the value of total interference is. Using the results presented in our previous works ([5], [22]), K can be obtained for the uplink of a DS-CDMA network with chip rate W , providing S services each with n_s users and activity factor ν_s , required bit-energy to interference plus noise spectral density, ρ_s , and bit-rate R_s as (see Appendix I)

$$K = \left(1 - (1 + f) \sum_{s=1}^S n_s \left(1 + \frac{B}{\rho_s \nu_s R_s} \right)^{-1} \right)^{-1}. \quad (4)$$

In (4), f is the ratio of the interference originated in the other cells to the interference generated in the home-cell which is a system parameter and is mainly a function of users' spatial distribution in the network coverage area [16]. As it is seen, for a given number of users in each service type (i.e., given n_s , $s = 1, \dots, S$), K in (4) is an increasing function of the users' activity. It is also observed in (4) that for a given set of services (i.e., given ν_s , $s = 1, \dots, S$), K is an increasing function of the number of users in the network coverage area.

In the case of a single cell single service network, or where there is no other-cell interference at the BS (i.e., $f = 0$, $S = 1$, $n_s = M$, $\nu_s = 1$), K is obtained from (4)

$$K = \frac{M_0}{M_0 - M}, \quad (5)$$

where M_0 is defined in (3). Note that K is only a function of the number of the active primary service users, M .

Consequently, similar to the primary users, for a secondary user that utilizes a B_c Hz sub-channel and is located in the coverage area of the primary network, the average interference imposed by the primary network at the receiver can be approximated by

$$I_{s|p} = \left(\tilde{K} - 1 \right) N_0 B_c, \quad (6)$$

where \tilde{K} is derived in [23].

In the following, for brevity of expositions, we assume that the secondary user accesses over only one sub-channel. If the secondary service access sub-channel i , a narrow-band interference, $I_{i,p|s}$, is correspondingly created at the receiver of the primary service (i.e., BS). The created interference $I_{i,p|s}$ may affect the performance of the primary system.

Here, we define *Perfect Underlay Spectrum Sharing*.

Definition (Perfect Underlay Spectrum Sharing): A perfect underlay spectrum sharing over sub-channel i does not change P_{out} in the primary network.

The following proposition gives a condition on $I_{i,p|s}$ that results in perfect underlay spectrum sharing.

Proposition 1: A perfect underlay spectrum sharing over sub-channel i exists if

$$I_{i,p|s} \leq N_0 B (M_0 - M - 1)^+ \triangleq Q_I, \quad (7)$$

where $(x)^+ = \max\{0, x\}$.

Proof: Without underlay spectrum sharing, E_b/I_0 constraint in (1) is $\frac{GQ_m}{\sum_{j \neq m} Q_j + N_0B} \geq \rho$, where the equality is held if

all the primary users j , $j = 1, \dots, M$ set their transmission powers such that the received power at the BS meets

$$Q_j = \frac{N_0 B}{M - M_0}, \quad M < M_0. \quad (8)$$

For the case of underlay spectrum sharing, due the secondary service transmission there is an extra narrow-band interference power $I_{i,p|s}$ imposed at the BS. However, the primary service is already able to provide service to $M_0 - 1$ primary users. Consequently, one can assign the maximum value of $I_{i,p|s}$, i.e., Q_I , such that the primary service acts similar to the case of serving $M_0 - 1$ users instead of actual M users as

$$\frac{N_0 B}{M_0 - M_0 + 1} = \frac{N_0 B + Q_I}{M_0 - M}. \quad (9)$$

Obtaining Q_I from (9) results in (7). ■

Proposition 1 indicates that if the secondary service activity causes $I_{i,p|s}$ to be greater than Q_I , then E_b/I_0 constraint is violated. Therefore, an outage incident is experienced. In such cases, the outage is a direct consequence of collision between the secondary service and the primary service transmissions.

For perfect underlay spectrum sharing, Proposition 1 suggests that the secondary service should adjust its transmission power to keep $I_{i,p|s} \leq Q_I$ at the receiver of the primary service. Therefore, the secondary transmitter requires the actual value of $I_{i,p|s}$ as well as g_{0i} to adjust its transmission power.

In practice, a direct signaling between two systems should be required for communicating the measured value of $I_{i,p|s}$ in the primary service, as well as g_{0i} , to the secondary service transmitter. Moreover, the primary network is either not able, or not particularly designed to measure and send back $I_{i,p|s}$ and g_{0i} to the secondary service. Therefore, the secondary service has no direct means to estimate these two parameters.

A. Outage and Collision

Here, for brevity, in the primary network the outage caused by collision is disjointed from the outages caused by other non-spectrum-sharing related effects. In other words, hereafter we consider collision as an outage exclusively caused by the secondary service activity. Therefore, the collision probability $P_{col(i)}$ is a metric that indicates the impact of underlay spectrum sharing over sub-channel i on the primary service. The acceptable level of collision probability, ξ , is defined by

$$P_{col(i)} = \Pr \{I_{i,p|s} > Q_I\} \leq \xi. \quad (10)$$

Generally, ξ is a system parameter set by either the primary service operator, or the spectrum sharing regulator. Thus, underlay spectrum sharing should act in a way that the collision probability in the primary network is always less than or equal to ξ . Hereafter, we refer to $\Pr \{I_{i,p|s} > Q_I\}$ as the collision probability.

B. Primary Network Power Control and Collision

In power controlled DS-CDMA, BS routinely reacts to the violations of E_b/I_0 constraint. Different parameters in the primary network may usually cause E_b/I_0 constraint to be violated. Having the secondary service active in the primary network, the power control mechanism also reacts to the violations of E_b/I_0 directly caused by the secondary

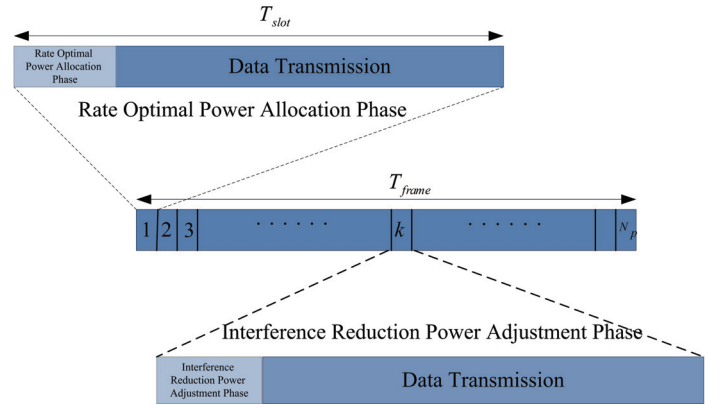


Fig. 2. Time is divided into frames (namely AMTPA frames), each frame consists of a number of slots, N_p . In the first slot of the frame, rate optimal power allocation phase is performed; in the consequent slots of the frame, interference reduction power adjustment phase is conducted.

service activity (i.e., correspondingly the violation of (7)). The reaction is to send appropriate power control commands to the affected primary service users. Thus, the power control commands, in a way, contain information on and about the interference level at the receiver of the primary service. Listening to the power control commands, the secondary user might be able to adjust its transmission power so that (7) is held. In the following, we present a method which exploits power control commands transmitted by the BS to the primary users, to adjust the power levels of the secondary users.

IV. AMTPA

The main objective of power allocation in the secondary service is to adjust the secondary service transmission power so that no quality degradation is imposed on the active primary service users. To achieve this objective, the values of $I_{i,p|s}$, and g_{0i} are required by the secondary service to set its transmission power, so that Proposition 1 is held. However, as it was mentioned previously, considering such signaling framework is not always attainable. To tackle this issue, we propose the Adaptive Multiple Time-scale Power Allocation (AMTPA) which exploits power control commands and does not require direct signaling between the primary and secondary systems.

AMTPA is a power allocation strategy adopted in the secondary service transmitter which consists of two phases in different time-scales (Fig. 2): *Rate Optimal Power Allocation* and *Interference Reduction Power Adjustment*. In the longer time scale, T_{frame} , (i.e., at the beginning of each frame), the rate optimal power allocation phase allocates the transmit power of the secondary service transmitter with the objective of maximizing the achievable capacity of the secondary channel and without considering Q_I in (7). The rate optimal power allocation phase may cause an increase in the received interference in the primary network; thus, results in decreasing E_b/I_0 and consequently increasing bit-error-rate, i.e., QoS degradation. Then, in the shorter time-scale, T_{slot} , the interference reduction power adjustment phase tries to reduce the QoS degradation through implementing the primary service interference constraint in (7) by adjusting the allocated transmit power of the secondary service.

A. Rate Optimal Power Allocation Phase

In the rate optimal power allocation phase which is over the longer time-scale, we assume that only shadowing effect is involved. This assumption can be justified by adopting Jensen's inequality as follows. The ergodic capacity of the secondary service over sub-channel i , C_{si} , is

$$C_{si} \leq B_c \mathbf{E}_{\text{shad}} \left\{ \log \left(1 + \frac{g_{1i} P_{si}}{\tilde{K} N_0 B_c} \right) \right\}, \quad (11)$$

where \mathbf{E}_{shad} indicate expectations over shadowing. Consequently, in (11), g_{1i} only contains shadowing effect. The optimal secondary service power allocation over sub-channel i in frame n is then obtained via water filling (WF) algorithm [24] as

$$P_{si,WF}(n) = \begin{cases} \frac{1}{\lambda_0} - \frac{\tilde{K} N_0 B_c}{g_{1i}(n)} & \text{if } g_{1i}(n) \geq \lambda_0 \tilde{K} N_0 B_c \\ 0 & \text{O.W.,} \end{cases} \quad (12)$$

where λ_0 is the Lagrangian coefficient adjusted to meet the transmission power constraint of the secondary service, $\bar{P}_{s,\max}$.

Finding sub-channel i itself, is an optimization problem which can be conducted with the objective which is a combination of imposing minimum interference over the primary service and achieving maximum capacity in the secondary system. Solving such optimization problem is elaborated in one of our previous works [25] in which both secondary and cross channels are required. However, since the cross channel is assumed to be unavailable, for brevity of expositions, we simply assume that the sub-channel in rate optimal power allocation phase is randomly selected. Obviously, adopting one of the optimal sub-channel selection scenarios in [25] a higher capacity is achieved in the rate optimal power allocation phase.

In AMTPA, the secondary service transmit power at the first slot of the n th frame, i.e., $P_{si}(n, 1)$, is set as the output of the rate optimal power allocation phase, i.e., $P_{si,WF}(n)$,

$$P_{si}(n, 1) = P_{si,WF}(n), \quad (13)$$

where $P_{si}(n, k)$ is the transmit power of the secondary service user at slot k in frame n of sub-channel i . Employing only the rate optimal power allocation phase, the secondary service transmit power during frame n remains equal to $P_{si,WF}(n)$. This allocated power imposes an equivalent narrow-band interference, i.e., $I_{i,p|s}(n)$, to the receiver of the primary service¹

$$I_{i,p|s}(n) = g_{0i}(n) P_{si,WF}(n). \quad (14)$$

Thus if $I_{i,p|s}(n) > Q_I$, a collision is occurred. The following proposition provides the probability of collision caused by the rate optimal power allocation phase in the primary network.

Proposition 2: If both g_{0i} and g_{1i} are log-normal shadowing with standard deviation σ_{dB} , the probability of collision caused by performing rate optimal power allocation phase over

¹Unless otherwise specified, it is assumed that the effect of distance dependent path-loss attenuation is scaled in the background noise power, and interference threshold Q_I , at the secondary, and primary service receivers, respectively.

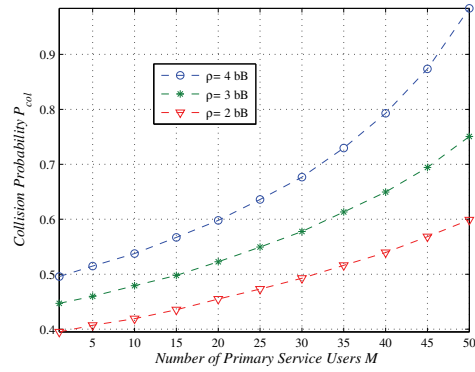


Fig. 3. Collision probability versus M in the rate optimal power allocation phase, ($G = 128$, $\bar{P}_{s,\max} = 1$ Watt, and $\sigma_{dB} = 8$ dB).

a sub-channel i is

$$P_{col(i)} = \frac{1}{2} \int_{\Phi} \operatorname{erfc} \left(\frac{\ln \left[\frac{Q_I}{\left(\frac{1}{\lambda_0} - \frac{\tilde{K} N_0 B_c}{\alpha} \right)} \right]}{\sqrt{2}\sigma} \right) f_{1i}(\alpha) d\alpha, \quad (15)$$

where, $\Phi \triangleq \left\{ \alpha \mid \alpha \geq \lambda_0 \tilde{K} N_0 B_c \right\}$, $\operatorname{erfc}(x) \triangleq \frac{2}{\sqrt{\pi}} \int_x^{\infty} e^{-t^2} dt$, $f_{1i}(x) = \frac{1}{x\sigma\sqrt{2\pi}} \exp \left(-\frac{(\ln x)^2}{2\sigma^2} \right)$, and $\sigma = 0.1 \log_{10} \sigma_{dB}$.

Proof: See Appendix II.

In Fig. 3, $P_{col(i)}$ is plotted versus number of the primary service users, M , for a randomly selected sub-channel i considering $G = 128$, where the standard deviation of log-normal shadowing, σ_{dB} , is 8 dB. According to Fig. 3, by increasing M , the probability of collision is increased since the amount of Q_I is decreased. Note that by increasing ρ , the probability of collision for a given M increases, because Q_I decreases.

The rate optimal power allocation phase does not consider Q_I and small-scale variations of wireless channel caused by fast fading. Therefore, the secondary service activity based on the allocated power in the rate optimal power allocation phase may cause collision and have destructive impact on the performance of the primary network. Interference reduction power adjustment phase in AMTPA has been designed to decrease the impact of rate optimal power allocation phase through adaptive adjustment of the transmission power in the shorter time-scales, i.e., T_{slot} .

B. Interference Reduction Power Adjustment Phase

Interference reduction power adjustment phase adaptively adjusts transmission power by exploiting the power control commands transmitted by the BS into the *probe user*. Probe user is a primary user m which is selected and associated with the secondary service either randomly, or by the primary network. In AMTPA, we set $P_{si}(n, 1) = P_{si,WF}(n)$, for slot 1 in frame n . Then, the interference reduction power adjustment phase, adjusts the transmit power based on $\Gamma_{pm}(n, k)$ so that the transmission power of the secondary service is decreased

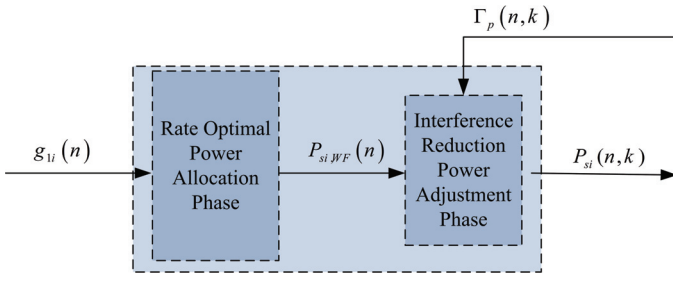


Fig. 4. AMTPA block diagram.

(increased) by Δ_s if $\Gamma_{pm}(n, k) = 1$ ($\Gamma_{pm}(n, k) = 0$), i.e.,

$$P_{si}(n, k) = \left(P_{si}(n, k-1) + \Delta_s (1 - 2\Gamma_{pm}(n, k)) \right)^+. \quad (16)$$

This is illustrated in Fig. 4.

Note that, after receiving power control commands, the secondary service takes action based on a pessimistic strategy presented in (16). In (16), it is implicitly assumed that any increase in the interference received by the primary user is exclusively caused by the secondary user. This argument emanates from the fact that the secondary service is not the licensee; thus should release the spectrum immediately, if it is required by the the primary service.

After applying the interference reduction power adjustment phase, E_b/I_0 requirement within frame n for user m , i.e., (1), is modified as

$$\frac{GQ_m}{\sum_{j \neq m} Q_j + I_{i,pl_s}(n, k) + N_0B} \geq \rho, \quad k = 1, \dots, N_p, \quad (17)$$

where $I_{i,pl_s}(n, k)$ is the narrow-band interference caused through applying (16) in the secondary transmitter. If for any reason (17) is violated at slot k , a power control command $\Gamma_{pm}(n, k+1) = 1$ is issued by the primary BS to the primary user to increase its transmission power by Δ_p Watts. After receiving this command, at the same time the secondary user reduces its transmission power by Δ_s Watts.

C. Modified-AMTPA

In the interference reduction power adjustment phase, setting $P_{si}(n, 1) = P_{si,WF}(n)$ causes the system history associated with $P_{si}(n-1, N_p)$, not to be incorporated into the interference reduction power adjustment phase in the subsequent slots. To incorporate the ‘‘system history’’ we propose the Modified-AMTPA, in which the interference reduction power adjustment phase is modified as

$$P_{si}(n, 1) = \min \{ P_{si,WF}(n), P_{si}(n-1, N_p) \}, \quad (18)$$

where $P_{si,WF}(n)$ is obtained by (12) and $P_{si}(n-1, N_p)$ is the transmitted power of the last slot of the previous frame i.e., $n-1$. Eq. (18) makes the power assignment for the first slot of each frame in favor of the primary service. Applying this modification, we expect the number of collisions and $P_{col(i)}$ to decrease. This claim is verified through simulations in Section VI.

D. Random-Modified-AMTPA

In addition to the operation of the secondary service, there might be other reasons for the BS to issue a power control

command, e.g., temporal variations of the wireless channel between the BS and the primary user and/or interference fluctuation in the primary network. However, in the interference reduction power adjustment phase we assume that the secondary service is the only reason for issuing a power control command. This is the reason we called the power allocation strategy in (16) a ‘‘pessimistic strategy’’ that particularly acts in favor of the primary network.

Moreover, in practice, to reduce delay and overhead, the power control commands are transmitted without error correcting codes. Therefore, approximately 10% of the power control command are received erroneous [26]. Let p_{error} be the probability of existence of error in the received power control command. The error can be modelled by multiplying $2\Gamma_{pm}(n, k) - 1$ by $1 - 2E_p(n, k)$, where $E_p(n, k)$ is 1 with probability p_{error} (error occurs), and 0 with probability $1 - p_{error}$ (error does not occur), otherwise.

To incorporate the above concerns, we propose the Random-Modified-AMTPA. In this method, (18) used for $P_{si}(n, 1)$, as it was in the Modified-AMTPA. However, the interference reduction power adjustment phase in the Random-Modified-AMTPA takes the power control command into consideration with a probability of p_{RM} , where p_{RM} is a system parameter that can be adjusted by experiment. In this approach, if $\Gamma_{pm}(n, k) = 1$ ($\Gamma_{pm}(n, k) = 0$), then the transmission power of the secondary service is decreased (increased) by the probability p_{RM} and remains the same with the probability of $1 - p_{RM}$.

E. Secondary Service Power Control

In cases where the secondary service has its own power control, the interference reduction power adjustment phase in (16) can be then conservatively modified as follows:

$$P_{si}(n, k) = \left(P_{si}(n, k-1) + \Delta_s (1 - 2\mathcal{T}(n, k)) \right)^+, \quad (19)$$

$$\mathcal{T}(n, k) = \Gamma_{(s)s}(n, k) \vee \Gamma_{pm}(n, k), \quad (20)$$

where ‘‘ \vee ’’ represents logical OR, and $\Gamma_{(s)s}(n, k)$ represents the power control command issued by the receiver of the secondary service in slot k of frame n , $\Gamma_{(s)s}(n, k) = 1$ ($\Gamma_{(s)s}(n, k) = 0$) corresponds to a decrease (increase) in the transmit power of the secondary user. It is worth mentioning that $\Gamma_{(s)s}(n, k) = 1$, means that the received signal at the secondary receiver has enough power to be decoded correctly.

F. Enforcing Primary Service Collision Probability Constraint

Utilizing Proposition 2, for a given collision probability in the primary network, ξ , we can obtain the corresponding Q_I . Using (14) the corresponding transmit power of the secondary service over sub-channel i in slot 1 of frame n is obtained using (13) as

$$P_{si}(n, 1) = \min \left\{ P_{si,WF}(n), \frac{Q_I}{g_{0i}(n)} \right\}. \quad (21)$$

We refer to this version as AMTPA with collision probability constraint enforcing (AMTPA-CPE). In cases where the secondary service knows the transmit power of the SCH channel (see Section V-A for a brief discussion on SCH) in the primary network, $g_{0i}(n)$ can be easily obtained assuming

reciprocity. Note that, in this case an estimate of g_{0i} is obtained in the shadowing time scale and still no signaling is required between the primary and secondary systems.

V. PRACTICAL CONSIDERATIONS

A. Primary Network

In DS-CDMA networks, a synchronization channel (SCH) is broadcasted by the BS to establish the coverage area of a cell [21]. The SCH also includes the cell-id and other necessary information for users to get connected to the network through the corresponding BS. In AMTPA, secondary users should utilize the information contained in the SCH, thus should be facilitated by the ability of receiving and decoding the SCH information.

B. The Probe User

The probe user is either selected randomly or associated with the secondary service by the primary network. It can be also set permanently by the primary network operator. Accordingly, the right to access the power control signaling of the probe user should be granted to the secondary service. Therefore, an entity in the primary network should provide the secondary user with the required information to receive and decode the power control command issued by the primary network to the probe user. Consequently, there should be a *cooperation framework* in higher layers between the two systems.

Actually, it could be possible for the primary network to consider a *virtual probe user*. In such scenario, the primary service BS generates a specific set of power control command for the secondary service based on the primary network situation. Transmitting power control commands acts as the signaling between the primary and secondary services. Therefore, in such case, AMTPA could act upon the received power control commands. We expect that in this case, AMPTA performance is significantly increased. However, one can say such scenario increases the complexity of the primary network and may not be attainable in practice.

C. Primary Network Stability and Feasibility

Practically in power controlled systems, there may be cases where, due to lack of enough radio resources, it is infeasible for the power control mechanism to satisfy the required E_b/I_0 for all active users at the same time. This situation is sometimes referred to as power control instability or infeasibility [27], [28]. In such cases, one solution is to adopt an appropriate removal strategy, which removes some active users, so that the power control becomes stable for the rest of the users [27], [28].

In AMTPA, the secondary service activity may cause the power control mechanism in the primary network to become instable. We note that the necessary condition for feasibility of the power control algorithm is that the spectral radius² of matrix $\tilde{\mathbf{H}}$ should be equal or less than 1, where the elements of $\tilde{\mathbf{H}}$ are the primary channel gains. If the power

of background noise at the BS is zero; therefore, the spectral radius of $\tilde{\mathbf{H}}$ is equal to 1. However, based on a nonzero noise power, an infinite transmission power of the primary service transmitters is required. Although in practice, the primary service transmitters have usually some constraints on their transmission powers. Therefore, for some finite number of the primary service users (e.g., $M < M_0$) the power control algorithm is feasible and a unique power vector can be obtained.

In case, where a secondary service accesses to the spectrum, matrix $\tilde{\mathbf{H}}$ remains unchanged. Therefore, the necessary condition for feasibility of the power control is not changed. However, increasing the interference at the primary service receiver may make the power control algorithm in the primary service infeasible. In this case, given that the secondary service is able to adjust its transmission power based on (7), we expect that for suitable Δ_s the primary service power control remains feasible, although optimal power vector is modified. Indeed, holding (7) enforces the DS-CDMA system with $M < M_0$ active primary service users to act as a system with M active users with the equivalent interference generated by M_0 active users. Since, the power control is already feasible for supporting up to M_0 active users, it remains feasible.

We can show that the primary network power control is feasible and using AMTPA-CPE, there is a region for Δ_s in which the power control in the primary network remains feasible. The obtained upper bound of Δ_s suggests us how to select the probe user. Using this upper bound, the BS can select the probe user based on the values of the channel power gains among the BS and the primary users, and eigenvalues of the primary users channel matrix.

As it was mentioned before, to maintain power control feasibility usually a removal strategy is adopted by the primary network. In such a case, a straightforward strategy would be a removal strategy so that the first priority for removal is the active secondary service users. In AMTPA we consider the following *secondary service dropping strategy* along with the removal strategy in the primary network. The secondary service transmitter stops transmitting if it receives k_D consequent reduce power control commands. System parameter k_D is selected by the primary service operator.

VI. SIMULATION RESULTS

We simulate a multiple secondary service spectrum sharing system including a primary cellular DS-CDMA network and a single secondary access point with 10 randomly selected sub-channels (see, Fig. 5). The wireless channel model consists of log-normal shadowing with a Rayleigh fast fading.

The primary network is a single cell DS-CDMA network with 2000 meter radius. Primary users are uniformly distributed in the coverage area. The system bandwidth is 5 MHz. A fast power control mechanism is performed in the primary network in which the BS transmits 1500 power control commands per second.

The secondary services utilized AMTPA for underlay spectrum sharing in the uplink spectrum of the primary network for downlink transmission between the secondary service access point and the secondary service users. The bandwidth of each sub-channel is 80 KHz. The secondary access-point has a

²The spectral radius of matrix \mathbf{A} with eigenvalues e_1, e_2, \dots , is defined as $\text{Spec}(\mathbf{A}) \triangleq \max_j e_j$.

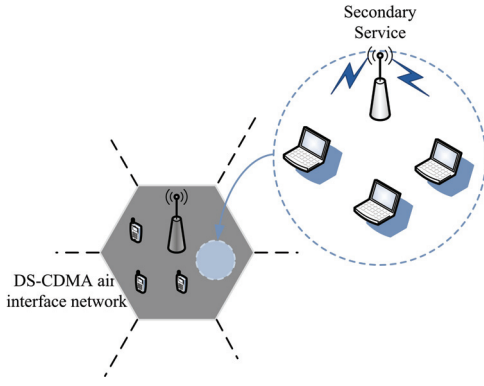


Fig. 5. The simulated spectrum sharing system with a DS-CDMA primary network and an OFDM-based secondary service access point.

TABLE I
SIMULATION PARAMETERS

Parameter	Value
N_p	32
DS-CDMA cell radius	2000 m
Secondary service cell radius	500 m
Spectrum Bandwidth	5 MHz
Standard dev. of log-normal shadowing	8 dB
Average of Rayleigh fading	1
Propagation loss exponent	4
Background noise power spectral density	-174.0 dBm/Hz
$\bar{P}_{s,max}$	1 Watt
G	128
Number of secondary users	10
Number of sub-channels N	64 (each 80 KHz)
Minimum required E_b/I_0 (primary network)	2 dB
M_0^\dagger	≈ 81
Primary user power constraint	1 Watt
Δ_p	0.1 Watt
Distance between the secondary and primary BSs	1000 m

[†]Parameter M_0 is obtained using (3).

circular coverage area with radius 500 meter. In each snapshot of the simulation the primary and secondary users are randomly located throughout the coverage area. The presented results is the average of the results from enough number of simulation snap-shots. Secondary service access point allocates its transmit power to the secondary service users in a way that they do not impose interference over each other. User m (the probe user) is randomly selected by the primary service access point and associated with a secondary service.

Simulation parameters are presented in Table I. The main performance measures are P_{col} (i.e., the probability of collision for selected sub-channels) and capacity of the secondary service which we study in the following considering different scenarios.

A. Impact of Δ_s

To study the impact of Δ_s , in the simulated system we obtain the collision probability of the primary service. The obtained collision probability is plotted versus Δ_s in Fig. 6 for different values of M . As it is seen, the probability of collision is decreased by increasing M . We elaborate on this later in this section. The observed pattern of the probability of collision variations is the same for different primary network loads (i.e., for different values of M). Note that $\Delta_s = 0$ is

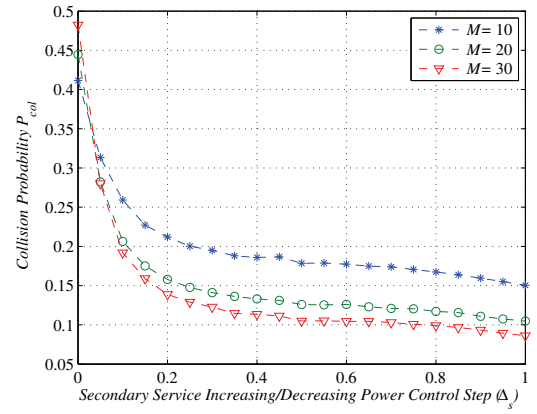


Fig. 6. Collision probability vs. Δ_s for $M = 10, 20, 30$.

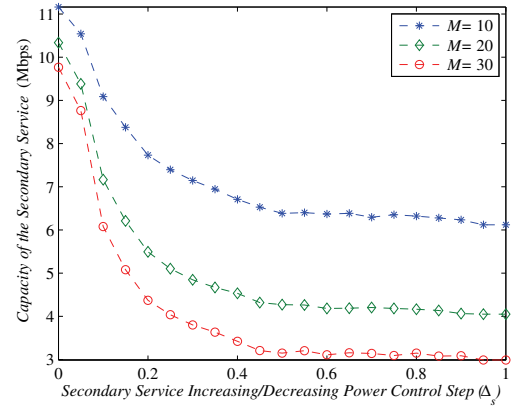


Fig. 7. Capacity of AMTPA vs. Δ_s for $M = 10, 20, 30$.

equivalent to AMTPA without interference reduction power adjustment phase.

By increasing Δ_s from 0 to $\Delta_s \approx 0.4\bar{P}_{s,max}$, probability of collision is significantly decreased. However, for $0.4\bar{P}_{s,max} \leq \Delta_s \leq \bar{P}_{s,max}$ it is seen that P_{col} is not approximately changed. This is due to the fact that for larger values of Δ_s the secondary service user actually stops transmission after receiving a couple of decreasing power control commands. Therefore, the collision probability is not affected.

We also study the impact of Δ_s on the capacity of the secondary service for different values of M in Fig. 7. As it is seen by increasing the number of the primary users, M , the capacity of the secondary service is decreased. Fig. 7 also indicates that by increasing Δ_s from 0 to $0.4\bar{P}_{s,max}$ the capacity of the secondary service is decreased. For $0.4\bar{P}_{s,max} \leq \Delta_s \leq \bar{P}_{s,max}$ the capacity of the secondary service is approximately constant. As it is seen in the two above simulations $\Delta_s = 0.4\bar{P}_{s,max}$ is appropriate for AMTPA. In the rest of our simulations we set $\Delta_s = 0.4\bar{P}_{s,max}$.

We compared the impact of Δ_s on AMTPA and Modified-AMTPA for a given value of $M = 10$ in Fig. 8. It is observed that the probability of collision for a given value of Δ_s in Modified-AMTPA is smaller than that of AMTPA. Furthermore, the Modified-AMTPA is *more sensitive to the choice of Δ_s* . Indeed, for $\Delta_s \approx 0.05\bar{P}_{s,max}$ the collision probability is minimum for Modified-AMTPA. For sufficiently large values of Δ_s , i.e., $\Delta_s \geq 0.4\bar{P}_{s,max}$, both of AMTPA and Modified-AMTPA follow the same patters and have the same collision

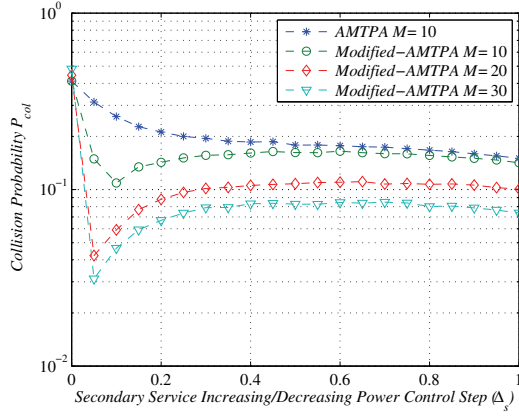


Fig. 8. Collision probability vs. Δ_s for AMTPA and Modified-AMTPA.

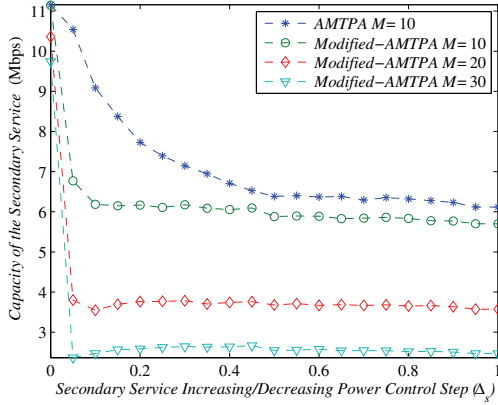


Fig. 9. Capacity of AMTPA and Modified-AMTPA vs. Δ_s .

performances. This is mainly due to the following. Although in AMTPA a higher power level transmission is allocated to the secondary service in the start slot of the new frame in the contrary of Modified-AMTPA, in both AMTPA and Modified-AMTPA the secondary service allocates a similar power level to the slots within each frame. Fig. 8 also presents the collision probability in Modified-AMTPA for different values of M . As it is seen, similar trend is observed in all cases; for $\Delta_s \approx 0.05\bar{P}_{s,max}$ the collision probability is minimum for Modified-AMTPA.

For small values of Δ_s , the Modified-AMTPA outperforms the AMTPA from the collision probability viewpoint. The rapid decrease observed in the collision probability for the Modified-AMTPA by increasing Δ_s from zero to approximately $0.1\bar{P}_{s,max}$, is due to the limiting effect of the secondary service transmission power in the first time slot of each frame. The limiting effect is because of the “min” operator which is implemented in the Modified-AMTPA. For moderate Δ_s , it is seen that the collision probability of Modified-AMTPA approaches to the AMTPA collision performance. This is because the fact that for moderate and large Δ_s in some cases the secondary service is allocated a higher power level at the last time slot of the previous frame, comparing to the power allocated in the first slot of the new frame based on water-filling. For large values of Δ_s both of the AMTPA and Modified-AMTPA behave similarly in collision probability perspective.

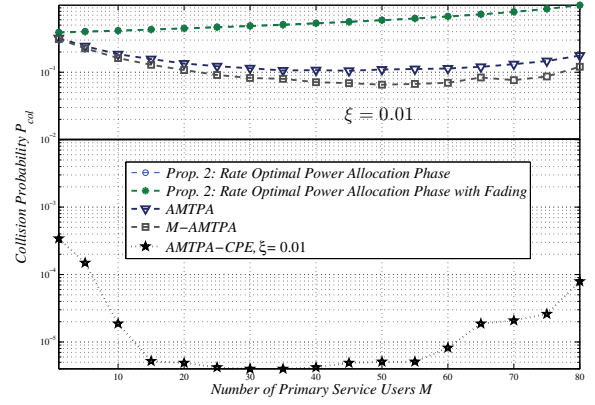


Fig. 10. Collision probability vs. M , for a system which only performs the rate optimal power allocation phase (AMTPA without power adjustment), and AMTPA variations.

In Fig. 9 we present the capacity of the secondary service versus Δ_s in AMTPA and Modified-AMTPA. As it is seen, the capacity of the Modified-AMTPA is smaller than that of AMTPA. However, similar to the pattern observed in Fig. 8, the Modified-AMTPA is more sensitive to the choice of Δ_s . Putting the results in Fig. 8 together with Fig. 9 we conclude that increasing M , decreases the collision probability and the secondary service capacity. Noting Fig. 8 and Fig. 9 shows that the Modified-AMTPA achieves better collision probability but worse capacity performance comparing to the AMTPA.

The above simulations shows that one can select Δ_s with different objectives such as decreasing the collision probability, or increasing the capacity.

B. Collision Probability

We study the collision probability in Fig. 10. We consider the obtained collision probability for rate optimal power allocation phase in Proposition 2 under shadowing channel model as a benchmark. Then, we compare it with the simulation results when fading is also considered for AMTPA and Modified-AMTPA.

First, we simulate the rate optimal power allocation phase for a system with both shadowing (which has been also assumed in derivations of Proposition 2) and fading. As it is seen in Fig. 10, simulation results follow the obtained collision probability in (15) closely. This is mainly due to the fact that large enough number of fading events (with a time scale close to slot duration), is observed within a frame (i.e., $N_p = 32$).

We then apply interference reduction power adjustment phase and conduct the simulation for the case where both fading and shadowing exist. Fig. 10 shows the corrective effect of the interference reduction power adjustment phase which significantly reduces the collision probability in comparison with AMTPA without interference reduction power adjustment phase. This result shows the effectiveness of interference reduction power adjustment phase in reducing the collision probability. We also observe that the gap between the collision probabilities in these two cases is increased by increasing M .

Fig. 10 also shows by increasing M from 1 to roughly 60 the collision probability in AMTPA is reduced. This is because of increasing the rate of the power decreasing

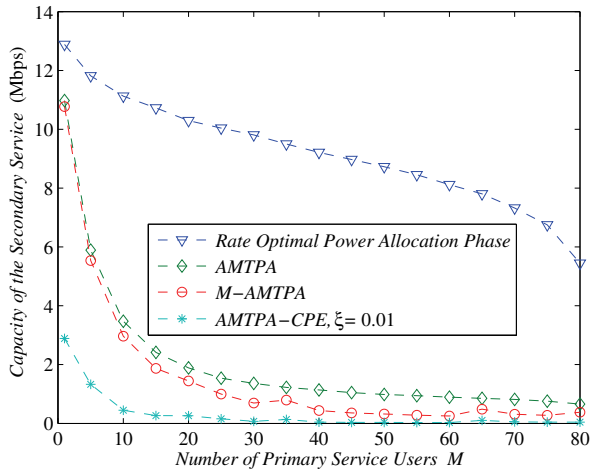


Fig. 11. Secondary service capacity vs. M , for a system which only performs the rate optimal power allocation phase (AMTPA without power adjustment) and a system which also performs the interference reduction power adjustment phase (AMTPA and its modifications).

commands issued by the primary service power control. For $60 \leq M \leq 80$, the collision probability is increased. In this case, on the one hand the amount of the interference threshold, Q_I , decreases and approaches to zero, and on the other hand, the rate of power reduction commands in the primary service power control is significantly increased. This results in rare power increase in the secondary service. However, even this rare event may result in collision because of small Q_I .

To show the impact of incorporating system history in AMTPA, we also compare the collision probability of AMTPA and that of Modified-AMTPA. Fig. 10 shows a slight improvement in the collision probability of Modified-AMTPA comparing to AMTPA.

We also simulate AMTPA with collision probability enforcing (AMTPA-CPE) for $\xi = 0.01$ as it was presented in Section IV-F. As it is seen in Fig. 10 using AMTPA-CPE, the actual collision probability always satisfies the acceptable collision probability by the primary network i.e., $\xi = 0.01$. This simulation shows that one can guarantee a certain level of collision probability in the primary network by conservatively selecting appropriate level of Q_I and implying it into the allocation phase. Note that in AMTPA-CPE, the target collision probability ξ is applied for the first time slot; thus, p_{col} is smaller than the targeted value, even in the first time slot. However, employing interference reduction power adjustment phase further decreases the collision probability.

This simulation indicates that providing a certain outage probability in the primary network is possible, however the cost is decreasing secondary service capacity. To investigate this, simulate a spectrum sharing system in which the secondary service only performs the rate optimal power allocation phase (i.e., AMTPA without power adjustment) and compare its capacity with a system which also performs the interference reduction power adjustment phase (AMTPA and its modifications). The results are presented in Fig. 11. As it is seen, in the case of the rate optimal power allocation phase, increasing M results in decreasing of the secondary service capacity. This is due to the fact that the received interference at the secondary service receivers imposed by the primary users is increased.

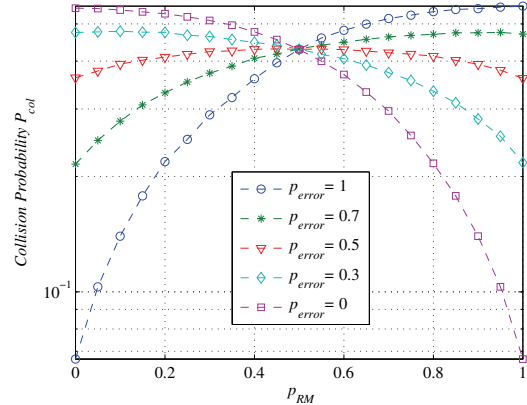


Fig. 12. Collision probability of Random-Modified-AMTPA vs. p_{RM} .

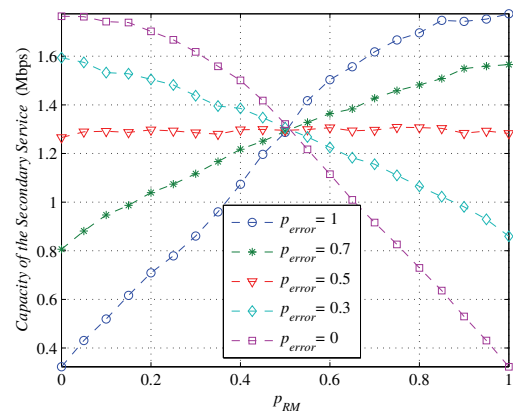


Fig. 13. Secondary service capacity of Random-Modified-AMTPA vs. p_{RM} .

Similar pattern is observed for AMTPA with the interference reduction power adjustment phase, Modified-AMTPA, and AMTPA-CPE. As it is seen, interference reduction power adjustment phase results in decreasing the secondary service capacity, comparing to the case where only the rate optimal power allocation phase is adopted. Furthermore, the secondary service capacity in Modified-AMTPA is smaller than that of in AMTPA and it is even lower in AMTPA-CPE. For $M \geq 40$ the capacity of the secondary service in AMTPA-CPE approaches zero since the interference threshold Q_I reduces to zero.

C. Impact of Error in Primary Power Control Commands

Here we also study the impact of error in power control commands. In Section IV-D we proposed a modification on AMTPA as Random-Modified-AMTPA. Fig. 12 shows the collision probability versus p_{RM} for $M = 30$ and various values of p_{error} , where p_{error} is the probability of existence of error in the received power control command.

For $p_{error} = 0$ the best collision performance is achieved when $p_{RM} = 1$. This means that if the power control commands are errorless, the secondary service should adjust transmit power based on the captured commands. For $p_{error} = 1$, the best collision performance is achieved for $p_{RM} = 0$.

It is also interesting to examine the impact of the error in the primary service power control on the capacity of the secondary service. Fig. 13 shows the secondary service capacity versus

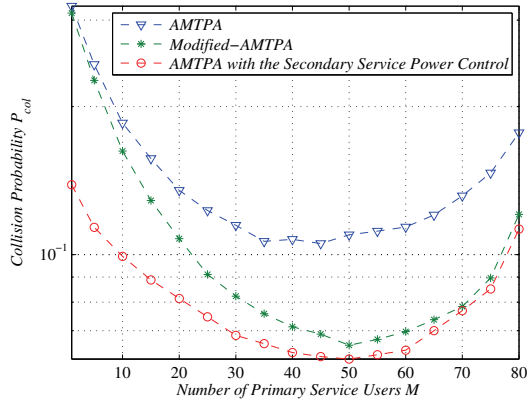


Fig. 14. Collision probability vs. M for AMTPA, Modified-AMTPA and AMTPA with the secondary service power control.

p_{RM} for $M = 30$ and various values of p_{error} . As it is also seen, in this case the trend of the capacity is similar to the observed collision probability behavior in Fig. 12.

D. Impact of Secondary Service Power Control

As mentioned before, the secondary service may also have a power control mechanism (see Section IV-E). Here we simulate a system in which the secondary service access point sends power control commands and the interference reduction power adjustment phase is based on (19). In Fig. 14 the collision probability is plotted versus M for AMTPA, Modified-AMTPA and AMTPA with secondary service power control. The required SINR at the secondary service receiver is 10 dB. It is seen that by considering the secondary power control command the collision probability is reduced comparing to AMTPA and Modified-AMTPA.

In AMTPA, the secondary service is able to increase its transmission power in some circumstances, while it is not really required to do this based on the received secondary service power control commands. Therefore, as Fig. 14 also indicates the collision probability is decreased. It must be noted that, this behavior is related to the value of the required SINR at the secondary service receiver.

E. Issuing Probability of an Increasing Power Control Command in the Primary Service

Issuing Probability of an Increasing Power Control Command (IPIPCC) in the primary system is related to the rate of violation of the minimum required SINR at the primary service. This violation may be caused by channel variations or co-channel interference. In this subsection we study the IPIPCC for different versions of AMTPA, system with no spectrum sharing, and a spectrum sharing system which only adopts rate optimal power allocation phase. The results have been plotted in Fig. 15.

The system with no spectrum sharing is assumed to be able to support M_0 users. For this system IPIPCC is upper bounded by 0.85 which is corresponding to fully loaded system with $M = M_0$. For the system which only adopts rate optimal power allocation phase Fig. 15 shows that the power control behavior is very similar to a fully loaded system even for very small values of M , i.e., $M \geq 5$. Consequently for a primary

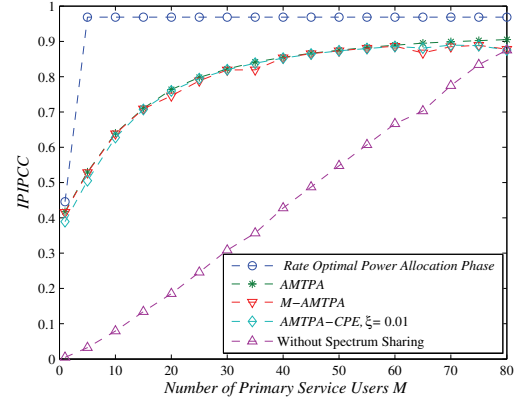


Fig. 15. IPIPCC vs. M for system with no spectrum sharing, AMTPA without interference reduction power adjustment, system which also performs the interference reduction power adjustment phase (AMTPA and its modifications).

user in this system experiencing an outage has high probability. It is seen that conducting the interference reduction power adjustment (i.e., utilizing AMTPA) significantly changes the primary service power control behavior, so that the system experiences same power control behavior in very higher values of M comparing to rate optimal power allocation system. This shows that using AMPTA the primary system with $M = 40$ users acts similar to a system with no spectrum sharing with 80 users. Similar pattern is also observed for Modified-AMTPA and AMTPA-CPE. This shows that AMTPA and its variations enable spectrum sharing in a power controlled primary service.

Looking to the plot from another perspective, reveals the fact that using rate optimal power allocation spectrum sharing makes the primary system to act as an overloaded network from the power control viewpoint. However, AMTPA and its variations never degrades the IPIPCC performance lower than that of a fully loaded system with no spectrum sharing.

VII. CONCLUDING REMARKS

In this paper, the Adaptive Multiple Time-Scale Power Allocation (AMTPA) method for underlay spectrum sharing in DS-CDMA wireless networks was presented. In AMTPA, the secondary service has access to the power control commands transmitted by the BS of the primary service. AMTPA manages the secondary service transmission power in two phases, each in different time-scales: the *rate optimal power allocation phase* and the *interference reduction power adjustment phase*. In the longer time-scale, the *rate optimal power allocation phase* adaptively allocates the transmission power of the secondary service transmitter using the instantaneous status of the secondary channel, to maximize the achieved capacity. The allocated power in the rate optimal power allocation phase causes interference to the primary service. Then, in the shorter time-scale, the *interference reduction power adjustment phase* exploits the power control commands to adjust its transmission power with the objective of reducing the impact of the secondary service transmission on the quality-of-service in the primary network. Taking into account some practical considerations, we also proposed modified versions of AMTPA that have a smaller probability of collision and is more robust against errors in reception of the power control commands.

Simulation results confirm our analysis and show that by using AMTPA, without using direct signaling between the two systems, spectrum sharing can be realized with a slight, but controllable level, of increase in the probability of collision. We also showed how one can set system parameters such as Δ_s , and p_{RM} so that the primary network collision probability constraint is satisfied.

APPENDIX I: OBTAINING EQUATION (4)

Consider the uplink of a DS-CDMA cellular network with B Hz bandwidth. For mobile user u_j , the average E_b/I_0 is assumed to be ρ_j which is

$$\rho_j = \frac{B}{\nu_j R_j} \frac{Q_j}{Q_{\text{total}} - Q_j}, \quad (22)$$

where ν_j , Q_j , and R_j are the activity factor, the average received power, and the bit rate of the j -th mobile user, respectively, and Q_{total} is the total received power of users plus noise at the BS. The load factor of mobile user j is denoted by η_j , where $Q_j = \eta_j Q_{\text{total}}$, and η_j indicates the contribution of user j in the total interference at the BS, Q_{total} . Using η_j in (22), it is easy to verify

$$\eta_j = \left(1 + \frac{B}{\rho_j \nu_j R_j}\right)^{-1}. \quad (23)$$

The total interference at the BS is divided into three different parts as $Q_{\text{total}} = Q_{\text{home}} + Q_{\text{other}} + N_0 B$, where Q_{home} is the interference generated by home cell users, Q_{other} is the interference generated by the users in the other cells. In line with the DS-CDMA literature (see, e.g., [22]), we also assume that $Q_{\text{other}} = f Q_{\text{home}}$. The parameter f defined as the ratio of other cell's interference to the home cell interference and called as network load factor. Consequently, we define η_{UP} as the uplink load factor which shows the contribution of the home cell and other cell interference into Q_{total} :

$$\eta_{\text{UP}} = \frac{Q_{\text{home}} + Q_{\text{other}}}{Q_{\text{total}}}. \quad (24)$$

We note that $\eta_{\text{UP}} = (1+f) \sum \eta_j$, then noting (23) for S different services each with n_s users, η_{UP} is obtained as

$$\eta_{\text{UP}} = (1+f) \sum_{s=1}^S n_s \eta_s = (1+f) \sum_{s=1}^S n_s \left(1 + \frac{B}{\rho_s \nu_s R_s}\right)^{-1}. \quad (25)$$

Rewriting Q_{total} as $Q_{\text{total}} = \eta_{\text{UP}} Q_{\text{total}} + N_0 B$, and replacing η_{UP} from (25), Q_{total} is obtained as

$$Q_{\text{total}} = \left(1 - (1+f) \sum_{s=1}^S n_s \left(1 + \frac{B}{\rho_s \nu_s R_s}\right)^{-1}\right)^{-1} N_0 B. \quad (26)$$

Thus, parameter K in $Q_{\text{total}} = K N_0 B$, is obtained as it is presented in (4). ■

APPENDIX II: PROOF OF PROPOSITION 2

Here we drop time index n for brevity. Using (14) and (10), the collision probability is obtained as

$$P_{\text{col}(i)} = \Pr \{g_{0i} P_{si,WF} > Q_I\}, \quad (27)$$

where $P_{si,WF}$ is given by (12). Therefore,

$$P_{\text{col}(i)} = \int_{\Phi} \Pr \left\{ g_{0i} P_{si,WF}(\alpha) > Q_I, \alpha \geq \lambda_0 \tilde{K} N_0 B_c \right\} \times f_{g_{1i}}(\alpha) d\alpha \quad (28)$$

$$= \int_{\Phi} \int_{\Psi} f_{g_{1i}}(\alpha) f_{g_{0i}}(\beta) d\alpha d\beta, \quad (29)$$

where $\Psi = \left\{ \beta \geq \frac{Q_I}{\frac{1}{\lambda_0} - \frac{K N_0 B_c}{\alpha}} \right\}$. In the above, g_{0i} and g_{1i} , $i = 1, \dots, N$, are i.i.d. random variables with log-normal distribution. For a log-normal random variable X with variance σ_{dB} , $X = e^Z$, where Z is a zero-mean Gaussian random variable with variance σ^2 , and $\sigma = 0.1 \log_{10} \sigma_{dB}$; therefore,

$$f_X(x) = \frac{1}{x\sigma\sqrt{2\pi}} \exp\left(-\frac{(\ln x)^2}{2\sigma^2}\right). \quad (30)$$

Substituting pdf of g_{0i} , i.e., (30), into (29) yields (15) and completes the proof. ■

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